Application Note

Application of the P8xC592 microcontroller with CAN-interface

Peter Buehring, Peter Hank

Product Concept & Application Laboratory Hamburg, F. R. Germany

Keywords

P8xC592 microcontroller, on-chip CAN controller, CPU load

 Report No
 : HKI/AN 91 014

 Date
 : 92-07-02

 Pages
 : 36

©Philips Export B.V.

All rights are reserved. Reproduction in whole or in part is prohibited without the prior written consent of the copyright owner. The information presented in this document does not form part of any quotation or contract, is believed to be accurate and reliable and may be changed without notice. No liability will be accepted by the publisher for any consequence of its use. Publication thereof does not convey nor imply any license under patent- or other industrial or intellectual property rights.

Philips Semiconductors





Summary:

The integrated circuit P8xC592 from Philips is a single-chip high performance microcontroller designed for use in automotive and general industrial applications. It is a high-end derivative of the 80C51 family CMOS microcontrollers.

In addition to the 80C51 standard features the P8xC592 provides a number of dedicated hardware features for sophisticated control applications. It includes timers, A/D converter, PWM output, UART and a CAN protocol controller (CAN = Controller Area Network) for serial communication.

The P8xC592 covers the complete CAN specification, offering important features such as multi-master serial communication capability with a high number of participating network nodes, programmable data transmission rate up to 1 Mbit/s and powerful error handling.

This technical publication puts special emphasis on CAN applications of the P8xC592. The application note provides a simple circuit example for a CAN module built with a P8xC592. Furthermore flowcharts are discussed to let the reader become familiar with the software aspects of CAN communication. A practical example shows that there is very little CPU load for the control of CAN communication.

Revision history:

91-10-16: 1st release

92-07-02: 2nd release: pages 6,11,31,34 revised

Philips Semiconductors

Table of Contents:

1.	Introduction
2.	P8xC592 Features
3.	Hardware Aspects
3.1 3.2 3.2.1	Main Circuitry5Transceiver: Physical Interface to the Transmission Medium6On-chip Transceiver Components7
4.	Software Aspects
$\begin{array}{c} 4.1 \\ 4.2 \\ 4.2.1 \\ 4.2.2 \\ 4.2.3 \\ 4.3 \\ 4.3.1 \\ 4.3.2 \\ 4.3.3 \\ 4.3.4 \\ 4.3.5 \\ 4.4 \\ 4.4.1 \\ 4.4.2 \\ 4.5 \\ 4.6 \\ 4.6.1 \\ 4.6.2 \end{array}$	CAN Registers8On-chip Interface Between CPU and CAN Controller8CAN Related Special Function Registers10Auto Address Increment11High Speed DMA11Basic Functions for CAN Communication12Initialization12Transmission12Reception14Sleep Mode14Comparator Switches15Control of CAN Communication16Interrupts16Polling18CPU Load for CAN - a Real Example19Calculation of the CPU Load for CAN Communication21CPU Load24Spare Time25
5.	Development Tools
5.1 5.1.1 5.1.2 5.2 5.2.1 5.2.2 5.2.2 5.3 5.4	Philips P8xC592 Evaluation Board27P8xC592 Evaluation Board Hardware28P8xC592 Evaluation Board Software28In-Circuit-Emulators28Philips Stand-alone Debug System (SDS)28ICE from Other Vendors29Programming Support for P87C592 (EPROM)29Debug Tools for CAN communication30
6.	References

Appendix

A.	Example program for investigation of CPU load	32
A.1	Receiver part	32
A.2	Transmitter part	35

Philips Semiconductors	Application Note	HKI/AN 91 014
Philips Semiconductors	Application Note	HKI/AN 91

1. Introduction

CAN (Controller Area Network) is an advanced serial communication protocol, which efficiently supports distributed real-time control with a very high safety level. CAN allows the flexible configuration of networks with different types of microprocessors and microcontrollers. Typical applications of CAN-based networks can be found in automotive and industrial environment:

- o Automotive Systems:
 - multiplex wiring (< 125kbit/s)
 - engine control, ABS etc. (up to 1 Mbit/s)

o Industrial Systems:

- field bus applications
- robotics, numeric machine control

Fig.1 shows a schematic network using the P8xC592 for distributed control applications. The kernel of each module is a Module Controller (CPU) that communicates via the Bus Controller (i.e. CAN controller) with the other modules. CPU and CAN controller can also be implemented on the very same chip, as it is in the P8xC592.

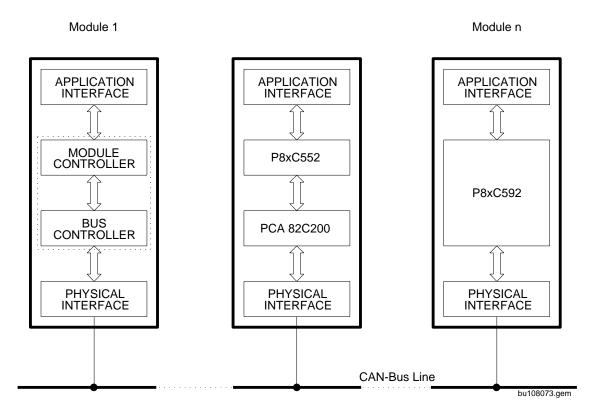


Fig. 1 Example of a "Controller Area Network"

The P8xC592 is a stand-alone high performance microcontroller designed for use in automotive and general industrial applications. In addition to the 80C51 standard features this device provides a number of dedicated hardware functions for these applications. It basically combines the functions of the well-known P8xC552 microcontroller [9] (without I^2C hardware) and the PCA 82C200 (Philips Standalone CAN-controller [3]), with some enhanced features.

Philips Semiconductors	Application Note	HKI/AN 91 014
------------------------	------------------	---------------

The CAN-part of the P8xC592 fulfills the complete CAN specification, to provide the following important features:

- o multi-master operation in a serial communication network with an unlimited number of active network nodes,
- o programmable data transmission rate, up to 1 Mbit/s,
- o very low probability of undetected errors, due to powerful error handling,
- o 40 m maximum distance between two bus nodes at a data transmission rate of 1 Mbit/s, lower transmission rates allow even longer distances,
- o guaranteed latency time supporting real-time applications.

This Application Note covers the CAN related items of P8xC592 applications. It describes a simple circuit example for a module in a CAN network. To apply and understand the application examples given in this document, the reader should be familiar with the Philips P8xC592 data sheet [2].

2. P8xC592 Features

Fig.2 shows the block diagram of the P83C592 (ROM) or the P87C592 (EPROM) versions. There is also a ROM-less version labeled P80C592 without on-chip program memory. Except the three blocks "CAN controller", "DMA" and "256 Bytes Aux. RAM", the blocks shown are not different from those which are present in other 80C51 derivatives like in the well-known Philips **P8xC552**. The use of the CPU, the memory access modes, the timers, A/D converter and PWM outputs have already been described in detail in the User Manual for the P8xC552 [9], which shall not be repeated in this application note again. Please refer to that document for information about these topics.

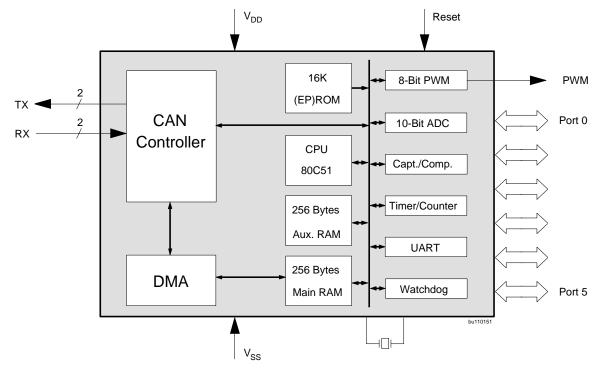


Fig. 2 Block diagram of the P8xC592

Philips SemiconductorsApplication NoteHKI/AN 91 014

The **CAN controller** on the P8xC592 is mainly an on-chip implementation of the Philips Stand-alone CAN-controller **PCA 82C200** [3]. The access to the internal CAN registers now is given via 4 Special Function Registers (**SFR**). The transfer of the messages between the CAN controller and the P8xC592's main RAM can be done in extremely short time by using Direct Memory Access (**DMA**). More information about this feature is given in chapter 4.2.3.

The P8xC592 has a 256 bytes **on-chip auxiliary RAM**, which is indirectly addressable in the same way as external data memory. This doubling of internal RAM capacity satisfies the rising demand for parameter storage in distributed control applications.

Application Note

Summarizing the P8xC592 is a single-chip 8-bit microcontroller featuring:

- o 80C51 CPU
- o CAN controller with DMA
- o 16K bytes ROM / EPROM
- o 2 * 256 bytes RAM
- o 10-bit A/D converter, 8 channels
- o 3 * 16 bit timer/counters
 - 4 capture registers
 - 3 comparators controlling 8 outputs
- o 6 * 8-bit I/O ports
- o 2 * PWM outputs
- o full-duplex UART
- o interrupt controller (15 sources)
- o watchdog timer

3. Hardware Aspects

This section covers hardware considerations of a CAN-based module in a network. The first part describes the minimum circuitry required for the P83C592 / P87C592. The second part gives suggestions for the connection of the controller to the CAN bus wires.

3.1 Main Circuitry

The P8xC592 is designed to work with a minimum of external components. Fig.3 shows the circuitry of a CAN node using the ROM- or EPROM- version P83C592 or P87C592. The only additional components that are required are a crystal (Q1) plus two small capacitors (C1, C2) to drive the on-chip oscillator, a reset circuit to provide the power-on RESET and a transceiver circuit (see section 3.2) for the connection to the bus wires.

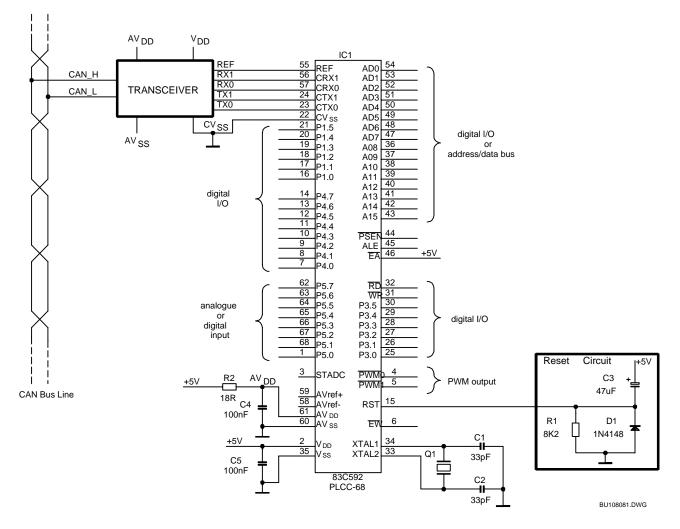


Fig. 3 P87C592 / P83C592 circuitry for a CAN application

Philips SemiconductorsApplication NoteHKI/AN 91 014

Power Supply

The 5V power supply is split on the different power input pins of the μ C. The main supply for the (digital part of the) CPU is fed to the **V**_{DD} / **V**_{SS} pins, buffered by capacitor C5.

It is recommended that the supply AV_{DD} / AV_{SS} for the analogue parts is derived from the main supply by the filter R2 / C4. The AV_{DD} / AV_{SS} supply is used to drive the on-chip A/D converter and the receiver section of the CAN controller; it shall also be used for optional biasing of the RX0 / RX1 pins in the off-chip transceiver.

The pin **CVss** is the ground pin belonging to the CAN on-chip transmitter output stages (cf. chapter 3.2.1), which shall also be used for optional ground potential of the off-chip transceiver.

Program Fetch

The schematics of fig.3 present the circuitry for microcontrollers with on-chip program memory (ROM or EPROM). Therefore the pin \overline{EA} is connected to +5V for program being fetched from internal memory. Note that the EA-pin is only read by the CPU during RESET, that means, switching it during program execution is not possible. This can give an additional protection against unauthorized copying of the on-chip program.

Reset Circuit

The schematics of fig.3 present a proposal for a discrete circuit that provides the necessary RESET signal to the CPU during power-up. The Reset Circuit can be replaced by a connection of the RST-pin to the power-on/power-fail reset output of the power supply, if available. Note that internal RESET conditions of the CPU (e.g. watchdog or recover from power-down) may cause the output of a short pulse on the RST pin, which the Reset Circuit has to tolerate.

3.2 Transceiver: Physical Interface to the Transmission Medium

The transceiver circuit performs the following functions:

- o it converts signals TX0, TX1 into the voltage levels for the bus wires,
- o it converts the voltage levels on the bus wires to be compatible with the CRX0, CRX1 inputs of the P8xC592.

How this connection actually is implemented, is application-specific. Depending on the requirements for the bus signals, the transceiver can be realized with more or less effort, the cheapest solution consists only of a couple of resistors, more expensive ones need some more discrete components or an extra IC. In principle the very same transceiver circuits can be used as for the stand-alone CAN controller PCA 82C200 [3]. Three examples for cheap discrete transceivers have been described in the Application Note [4] for the PCA 82C200, for instance.

In order to get to an international standard for CAN communication (including the transceiver), the International Standardization Organization (ISO) has prepared two standards, one for lower bit rates (up to 125 kbit/s) [6], one for higher ones [7]. They also define the electrical parameters for the transceiver. Application hints for implementation of an ISO-compatible transceiver circuit are given in [10].

Philips SemiconductorsApplication NoteHKI/AN 91 0	14	
---	----	--

3.2.1 On-chip Transceiver Components

The off-chip transceiver circuit connects the bus wires to the on-chip transceiver components. The on-chip transceiver provides the transmitter output stage and the receiver input comparator, see fig.4.

Transmitter

The transmitter provides two output lines (CTX0, CTX1); their characteristics can be programmed individually with the OUTPUT CONTROL register. Each line can work as open-drain or open-source or push-pull output, with positive or negative polarity. Thus the transmitter is well prepared to drive any kind of differential bus lines.

The connections of the transmitter outputs to the bus wires have to be done in such a way, that the resulting level on the bus is:

- o recessive level, when all nodes transmit a logical 1 or do not transmit,
- o dominant level, when one or more nodes transmit a logical 0.

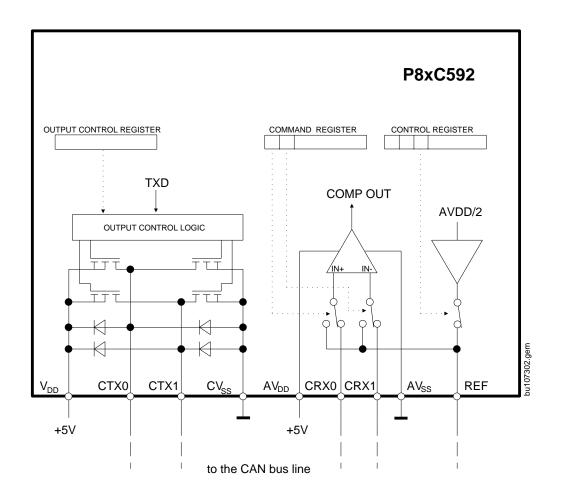


Fig. 4 Structure of the on-chip CAN transceiver

Philips SemiconductorsApplication NoteHKI/AN 91 014

Receiver

The on-chip receiver is a differential input comparator with the input lines CRX0 and CRX1. It decodes

- o the recessive level to a logical 1, when the voltage at the CRX0-pin is higher than at the CRX1-pin,
- o the dominant level to a logical 0, when the voltage at the CRX0-pin is lower than at the CRX1-pin.

Instead of decoding the **differential** bus signal, the comparator inputs can also be switched by software to decode the signal of only one of the two bus lines, comparing it to the reference voltage at the REF pin. The reference voltage can either be provided by the internal on-chip voltage generator (= default), or, if desired, it is fed to the REF pin by an external source, with the internal source being switched off by software. The use of **single wire** decoding is beneficial to continue communication, when one of the two bus lines has a wiring failure (open or short circuit), please refer to [8].

4. Software Aspects

4.1 CAN Registers

The P8xC592's on-chip CAN controller is a full implementation of the CAN protocol. It contains all necessary features required for a high performance communication protocol.

The CAN controller appears to the CPU as a memory- mapped I/O device which is arranged as Control Segment, Transmit (Tx) Buffer and Receive (Rx) Buffer (see fig.5).

Exchange of status, control and command signals between the CPU and the CAN controller is done by the control segment which contains 10 bytes. It is programmed during initialization for configuration of the communication parameters. Additionally the CPU controls the CAN communication via this segment.

Data to be transmitted are loaded into the Transmit Buffer by the CPU. The buffer contains the descriptor (Identifier, RTR-bit and DLC) and up to eight data bytes.

After a successful reception messages are read from the Receive Buffer. It consists of two 10 byte memories which are alternatively used to store messages. The CPU can process one message while another is being received.

4.2 On-chip Interface Between CPU and CAN Controller

To access the described CAN controller registers, four special function registers CANADR, CANDAT, CANCON and CANSTA are implemented. All CAN registers of the Control Segment which are used during communication now appear as being directly addressable as they are represented by these four special function registers. The Status Register represented by CANSTA now is even bit-addressable. With the help of these registers and the DMA logic, data transfer between Transmit/Receive Buffers and internal main RAM is also done very effectively (see fig. 6).

Philips Semiconductors

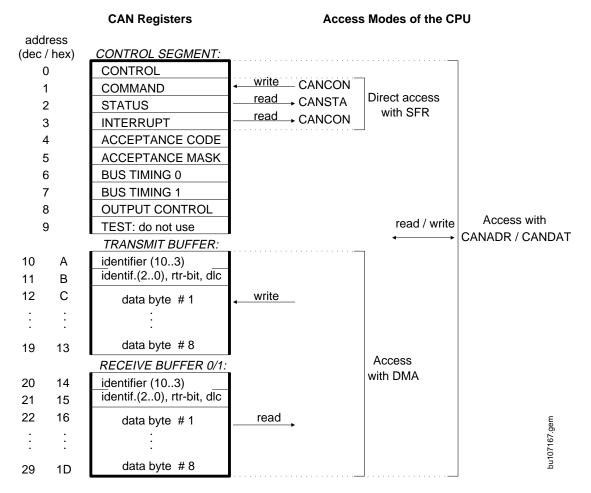


Fig. 5 CAN register mapping

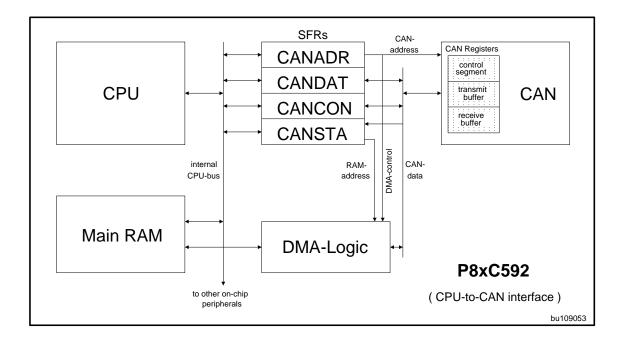


Fig. 6 Block diagram of the CPU - to - CAN interface

Philips Semiconductors	Application Note	HKI/AN 91 014
------------------------	------------------	---------------

4.2.1 CAN Related Special Function Registers

Table 1 shows the special function register arrangement. Note that CANCON and CANSTA have different meanings for a read and write access.

SFR	ADR	ACS	MSB 7	6	5	4	3	2	1	LSB 0
CANADR	D8H	R/W	DMA	reserved	AutoInc	CANA4	CANA3	CANA2	CANA1	CANA0
CANDAT	DAH	R/W	CAND7	CAND6	CAND5	CAND4	CAND3	CAND2	CAND1	CAND0
CANCON	D9H	R	reserved	reserved	reserved	Wake up Int	Overrun Int	Error Int	Transmit Int	Receive Int
		w	RX0 active	RX1 active	Wake up Mode	Sleep	Clear Overrun	Release Rx Buffer	Abort Transm.	Transmit Request
CANSTA	D8H	R	Bus Status	Error Status	Transmit Status	Receive Status	TX compl Status	Tx Buffer Access	Data Overrun	Rx Buffer Status
		w	RAMA7	RAMA6	RAMA5	RAM4	RAMA3	RAMA2	RAMA2	RAMA1

 Table 1: CAN Special Function Registers

HA107221

CANADR

With the least significant bits CANA4...CANA0 each of the CAN controller internal registers can be addressed by a write access to CANADR. Reading or modification of the particular CAN register then is done by reading or writing to CANDAT. CANADR is implemented as a read/write register which also contains control bits to select auto increment addressing and to start a DMA transfer.

CANDAT

When reading or writing CANDAT, access to the CAN registers addressed by CANADR is possible. The way of accessing CAN registers via CANADR/CANDAT normally is only necessary for the registers Acceptance Code, Acceptance Mask, Bus Timing 0, Bus Timing 1, Output Control, which the CPU has to access during initialization only.

CANCON

CANCON is a register with different meaning for write and read operation. Writing to CANCON is a direct access to the command register and reading from it is a direct access to the interrupt register.

CANSTA

The bit addressable register allows a direct read access to the Status Register of the CAN Controller. Writing to CANSTA sets the address of the on-chip main RAM for a subsequent DMA transfer.

Philips Semiconductors	
------------------------	--

4.2.2 Auto Address Increment

Fast reading and writing of consecutive CAN Controller internal registers is possible by setting the AutoInc bit of CANADR and the concerning register address simultaneously. A first access to CANDAT refers to the register specified in CANADR. After any read or write access to CANDAT the contents of CANADR are incremented automatically in this mode.

Incrementing CANADR beyond XX111111B resets the AutoInc bit automatically to XX000000B.

4.2.3 High Speed DMA

The DMA logic provides a very fast transfer of complete messages between receive/transmit buffers and internal data memory (Main RAM) within 2 instruction cycles. The transfer process operates in the background and therefore the CPU can continue with the next instruction. However, an access to the Main RAM or to the CAN special function registers is not allowed during this time.

A DMA transfer action is achieved by first writing the RAM address into CANSTA and then writing the Tx- or Rx-Buffer address and the DMA bit simultaneously into CANADR.

The DMA automatically recognizes the transfer direction (see table 2). For the Rx Buffer DMA there is the option to copy a whole message or only a part of it.

CANADR	subject to transfer	transfer direction
8AH	whole message	RAM -> Tx Buffer
94H 95H	whole message whole message without first byte	Rx Buffer -> RAM
96H	data bytes only *)	
97H	last data bytes *)	
	•	
9DH	only the last data byte *)	

*) data bytes are copied as far as they are available (DLC!)

Table 2: DMA Modes

Setting the DMA bit causes an automatic evaluation of the data length and subsequent transfer. For a Tx Buffer DMA transfer the data length is always expected at "RAM address + 1".

After the DMA transfer has been finished, the DMA bit and the AutoInc bit are reset.

4.3 Basic Functions for CAN Communication

4.3.1 Initialization

During Initialization the P8xC592's CAN Controller is configured according to the required communication parameters. Before setting the registers in the Control Segment (see fig.5) first the Reset Request bit in the Control Register has to be set to "present" (high). An initialization procedure should include definitions of the following items:

- o Acceptance Filter
- o Bus Timing
- o Output Mode
- o Interrupts
- o Reference Voltage Mode

A flow chart of the initialization procedure is shown in table 4.

When the configuration of the P8xC592 is finished the Reset Request bit has to be set to "absent" to allow receive- or transmit operations.

4.3.2 Transmission

The CPU may write a message into the transmit buffer whenever the transmit buffer access bit (Status Register) is set. After writing the message to the transmit buffer and setting the transmission request bit high (Command Register), the CAN Controller begins with the transmission process.

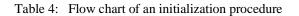
If the CPU cannot access the transmit buffer because a previously requested message is still waiting for transmission, it is possible to abort the current process. A transmission already in progress is not stopped.

if (status bit TRANSMIT BUFFE)	R ACCESS = "released") -+elseelse
write message into TRANSMIT BUFFER +	<pre> if (<high be="" message="" priority="" to="" transmitted="">) +thenthenelse+ set command bit ABORT TRANSMISSION:= "present" </high></pre>
REQUEST := "present" 	<pre>++ (* the next transmission is delayed until TRANSMIT BUFFER ACCESS = "released" is signalled by a TRANSMIT INTERRUPT or by polling the STATUS REGISTER *)</pre>

Table 3: Flow chart of a transmission procedure

Philips SemiconductorsApplication NoteHKI/AN 91 014

_____ _____ (* start initialization; write CONTROL REGISTER *) - set RESET_REQUEST := "present" - set TEST_MODE := "disabled" _____ (* write acceptance filter *) - write ACCEPTANCE CODE REGISTER - write ACCEPTANCE MASK REGISTER (* example: contents of ACCEPTANCE CODE REGISTER = 01110010B contents of ACCEPTANCE MASK REGISTER = 00111000B allows for messages with following IDs: 01xxx010xxxB *) _____ _____ (* define bus timing (baud rate on the CAN Bus) *) - write BUS TIMING 0 REGISTER - write BUS TIMING 1 REGISTER (* example: contents of BUS TIMING 0 REGISTER = 10001001B contents of BUS TIMING 1 REGISTER = 11101011B results in: = $(9 + 1) * 2 * t_{OSC}$ (with $f_{OSC} = 16$ MHz: 1.25 us) tscl = (2 + 1) * t_{SCL} (3.75 us) tsJW = 1 (three samples / bit taken) SAM tsync = 1 * t_{SCL} (1.25 us) $t_{TSEG2} = (6 + 1) * t_{SCL}$ (8.75 us) (15.00 us) $t_{TSEG1} = (11 + 1) * t_{SCL}$ = $t_{SYNCSEG} + t_{TSEG1} + t_{TSEG2} = 20 * t_{SCL}$ (25.00 us) tBIT variation of bit time due to resynchronization: (21.25 us) tBITmin = tBIT - tSJW = 17 tSCL tBITMAX = tBIT + tSJW = 23 tSCL (28.75 us) *****) _____ _____ (* define bus driver characteristics *) - write OUTPUT CONTROL REGISTER (* example: contents of OUTPUT CONTROL REGISTER = 10101010B results in: output mode: normal 1 (bit sequence on TXO and TX1 pins) TX0 configuration: bit-state 0 (1) -> output-state low (float) TX1 configuration: bit-state 0 (1) -> output-state high (float) *) _____ (* end initialization; write to CONTROL REGISTER *) (* bits 4 .. 1 *) - enable interrupt sources (* bit 5 *) - define mode of reference voltage (* SYNCH; bit 6 *) - define the resynchronization mode *) (* bit 0 - set RESET REQUEST := "absent" ------(* = start of comment *) = end of comment





4.3.3 Reception

Whenever the status bit 'Receive Buffer Status' is set, a new message is available in the Rx Buffer. Polling the Status Register enables the information of new message data in the Rx Buffer to be coordinated and controlled by the CPU.

When the Receive Interrupt is enabled the according bit in the interrupt register is set simultaneously with the Receive Buffer Status bit and the interrupt service can start with reading of message data from the Rx Buffer.

After Reading the buffer contents the CPU has to release this buffer by setting the Release Receive Buffer command bit. This may result in another message becoming immediately available.

+								+
Iread	lmessage	e fro	om RECEIN	/E BUFFER	ર			1
	5							ا + +
								+
lser	command	DIC	RELEASE	RECEIVE	BOLLEK	• =	"released"	I
+								+

Table 5: Flow chart of a reception procedure

4.3.4 Sleep Mode

If the Sleep Bit of the Command Register is set high the CAN controller enters the Sleep Mode, requiring that there is no bus activity and no interrupt is pending. A network enters the 'Sleep Mode' when all its nodes enter the Sleep Mode.

The CAN Controller wakes up after setting the Sleep bit of the Command Register low or when there is any bus activity. Upon wake up a wake up interrupt is generated. It is not necessary to enable the wake up interrupt.

After the CAN Controller entered Sleep Mode the CPU may be set into the 'Power Down Mode' (oscillator stopped). Upon wake up, the oscillator is started again. The wake up interrupt will wake up a CPU from 'Power Down Mode' by generating a Reset pulse, if the SIO1 (CAN) interrupt source was enabled. A CAN Controller that wakes up because of bus activity is not able to receive this message until it detects a bus free signal.

Reading of the Sleep bit reflects the status of the CAN Controller.

+	+
Set command bit SLEEP := "sleep"	
+	+

Table 6: Flow chart of a "go to sleep" procedure

Philips Semiconductors

Application Note

4.3.5 Comparator Switches

Data communication between the nodes of a network is disturbed or stopped when the bus wires are short-circuited (global failures) or interrupted (local failures). With the P8xC592 precautions have been taken to continue data communication despite of a bus failure.

With the help of the comparator switches, implemented in the CAN Controller, it is possible to tolerate the following single wiring failures:

- o one of the bus lines is short-circuited with VBAT or GND
- o one bus line is short-circuited against the other
- o interruption of one of the bus lines

The structure of the on-chip CAN Transceiver in fig.4 shows that in case of a short circuit it is possible to disconnect a blocked bus wire from the CAN receivers of all network nodes and replace this faulty bus line potential by a local reference level at one of the comparator inputs. In some cases it might be necessary to disconnect the bias voltage of one bus wire from the power supply terminal, additionally, to continue with communication.

Bus line interruptions are solved similar to the short circuit failures by replacing the potential of one bus line with a local reference potential.

The positions of the switches are defined via the Command Register bits RX0A and RX1A (see table 7). Reading of RX0A/RX1A reflects the status of the switches.

RX0A	RX1A	Result
0	0	unchanged
0	1	IN+=REF, $IN-=CRX1$
1	0	IN+=CRX0, $IN-=REF$
1	1	IN+=CRX0, $IN-=CRX1$

Table 7: Results of setting RX0A/RX1A

With the help of the described comparator switches three configurations can be selected:

- o two wire communication (differential signal)
- o single wire communication using CAN_H
- o single wire communication using CAN_L

Thereby a test of the bus wires can be implemented easily by software: if single wire communication is possible in the network with each of the bus lines as well as differential communication, then there is no wiring failure. Otherwise the faulty wire is detected (**diagnosis**) and the network may continue to communicate on the other wire till the faulty wire is repaired.

For more details on CAN Bus failure management please refer to [8].

Philips Semiconductors Application Note HKI/AN 91 014

4.4 Control of CAN Communication

Information about the reception of messages, the completion of a transmission, data overrun, error status etc. can be signalled to the CPU either with interrupt processing, described in chapter 4.4.1 or with polling, described in chapter 4.4.2. It can be of advantage to mix both, polling and interrupt control. Note, that the wake up event can only be processed by interrupt.

4.4.1 Interrupts

The P8xC592 controls five different CAN related interrupt sources. These are the Receive-, Transmit-, Overrun-, Error- and the Wake up Interrupt. All interrupts can be enabled/disabled via the Control Register except of the Wake up Interrupt which is always enabled. Setting of the interrupt sources is usually done during initialization (see also table 4). Do not forget to enable the CPU's SIO1 interrupt (by ES1 bit) and the global interrupt (by EA bit).

Upon the occurrence of one or more (enabled) interrupts the according bits are set in the **Interrupt Register** and a CAN interrupt (SIO1) for the CPU is performed. After reading this register, that appears to the CPU as a read only memory, all bits are reset by the CAN Controller. Therefore, the Interrupt Register should be stored in a bitaddressable part of the CPU memory for later interrupt evaluation.

Table 8 shows the procedure for an Interrupt Register data evaluation. In this example it is assumed that all interrupt sources are enabled to present a complete description.

The following summary gives a short overview of the CAN related interrupts and their occurrences.

Receive Interrupt:	If a new message is available in the Receive Buffer, the Receive Interrupt is given.
Transmit Interrupt:	A Transmit Interrupt is signalled, when a transmission has been completed or aborted.
Error Interrupt:	If the Bus Status or the Error Status changes, an Error Interrupt is given.
Overrun Interrupt:	When both Receive Buffers contain a message and the first byte of the next message should be stored, the Overrun Interrupt is signalled.
Wake up Interrupt:	Whenever the Sleep Mode is left, a Wake up Interrupt is signalled.

When CAN communication is controlled partly in the interrupt service routine and partly in the main program, care has to be taken about the use of CANADR/CANDAT (indirect addressing of the CAN Registers) or CANADR/CANSTA (use of DMA), respectively. If they are used as well in main program level as in interrupt level, it might be necessary to save the address of CANADR (push/pop) during interrupt processing, or disable the CAN interrupt during DMA transfer at main level, respectively. For the latter, if required, the interrupt has to be disabled right before the RAM address is written into CANSTA, and can be re-enabled just after the start of the DMA transfer.

|read INTERRUPT REGISTER and store in the bitaddressable part of the CPU memory| if (WAKE UP INTERRUPT = "set") -----then----else---if (<CPU state> = "just being awoken") +-----else-----then-------+----else-----else-------(* the network was awoken by some (* the just issued GOTO SLEEP CAN Bus activity command was not successful *) *) Perform appropriate action Perform appropriate network wake up activities _____+ if (TRANSMIT INTERRUPT = "set") +-----then-------+--else--+ (* the TRANSMIT BUFFER is released *) A next message may be written into the TRANSMIT BUFFER if (<using the ABORT TRANSMISSION command>) -----then------+----else-if (TRANSMISSION COMPLETE STATUS = "complete") | (* last requested +-----else-----+----else-----+ transmission has (* last requested (* last requested been completed transmission has | transmission has not |*) been completed, | been completed; | no ABORT TRANSMISS- the ABORT ION command has | TRANSMISSION command been executed *) | was successful *) if (**RECEIVE INTERRUPT** = "set") ----then-----+-else-|read RECEIVE BUFFER into CPU memory _____ _____+ |set command bit RELEASE RECEIVE BUFFER := "released" if (**OVERRUN INTERRUPT** = "set") -----then------+-else-an application may use this information to change its |behaviour (e.g. faster reaction on a RECEIVE INTERRUPT) |set command bit CLEAR OVERRUN := "clear" if (ERROR INTERRUPT = "set") -----then-----_____ ----+-else--if (BUS STATUS = "on-bus ") -----then----------else------(* P8xC592 takes part in bus (* P8xC592 does not take part in activities bus activities *) . | *) | if (<restart CAN required>) if (ERROR STATUS = "ok") -----then----+----else-----then-----then-----else-----(* the CAN Bus (* the CAN Bus |set control bit (* is presently | is presently |RESET REQUEST := not severely | severely |"absent" CAN hardware is left in disturbed | (* now the P8xC592 "off-bus" and disturbed waits for 128 * 11 "reset" state *) İ*) (*) consecutive recessive bits |perform applibefore going |cation-specif. "on-bus" again default-mode *****) operation ____+

 Table 8:
 Flow chart of Interrupt Register data evaluation

Philips Semiconductors	Application Note	HKI/AN 91 014
------------------------	------------------	---------------

4.4.2 Polling

Analysing the contents of the Status Register, as shown in table 9, is very similar to analysing the contents of the Interrupt Register. With the following overview the concerning Status bits are described:

Receive Buffer Status:	This bit is set when a new message is available.
Transmit Buffer Access:	The CPU may write a message to the Tx Buffer when this bit is set.
Error Status:	The Error Status bit is set when at least one of the Error Counters has reached the warning limit.
Data Overrun:	When both Receive Buffers are full and the first byte of a new message should

be stored, this bit is set.

_____ _____ if (**TRANSMIT BUFFER ACCESS** = "released") a next message may be written into the TRANSMIT BUFFER if (<using the ABORT TRANSMISSION command>) -----then-----else-----if (TRANSMISSION COMPLETE STATUS = "complete") |(* last requested transmission has (* last requested |(* last requested been completed transmission has | transmission has not i*) been completed, been completed due to no ABORT TRANS- an ABORT TRANSMISSION MISSION command command has been exec. *) *) ----+---------+ if (RECEIVE BUFFER STATUS = "full") -----+-else--+ _____ --then |read RECEIVE BUFFER into CPU memory +----set command bit RELEASE RECEIVE BUFFER := "released" if (DATA OVERRUN = "overrun") an application may use this information to change its |behaviour (e.g. faster reaction on a RECEIVE INTERRUPT) |set command bit CLEAR OVERRUN := "clear" +-----_____ if (BUS STATUS = "on-bus") (* P8xC592 takes part in bus |(* P8xC592 does not take part in bus activities activities İ*) *) if (**ERROR STATUS** = "ok") if (<restart CAN hardware required>) ----then-----+----else-----+----then-----++----else-----(* the CAN-bus is (* the CAN-bus is |set control bit RESET |perform presently severely presently REQUEST := "absent" |applicationnot severely specific disturbed (* now the P8xC592 disturbed default-mode waits for 128 * 11 . |*) *) operation consecutive some application- recessive bits specific actions before going may be necessary "on-bus" again (*)

 Table 9:
 Flow chart of Status Register data evaluation (polling)

Philips Semiconductors	Application Note	HKI/AN 91 014
------------------------	------------------	---------------

4.5 CPU Load for CAN - a Real Example

In a CAN system data transfer can be done with a very high level of safety with the CAN controller part of the P8xC592. The CPU has only the task of post-processing the received data and pre-processing the data to be transmitted. Communication between CPU and CAN controller is easily done via the Special Function Registers described in chapter 4.2.1. Using the DMA transfer facility, data exchange between CPU and Rx/Tx buffer is very effective.

That means that the resulting CPU load for CAN communication is very low, as the analysis of the following communication example proves. Simulations with "NetSim" (a simulator for CAN communication, see chapter 5.4) have been done with a real communication example of 7 nodes. At a data rate of 500kbit/s, 12 different messages M are transferred on the bus. Fig.7 shows the structure of the network. M3_2 means: 2nd type of message that is transmitted by node number 3, for instance.

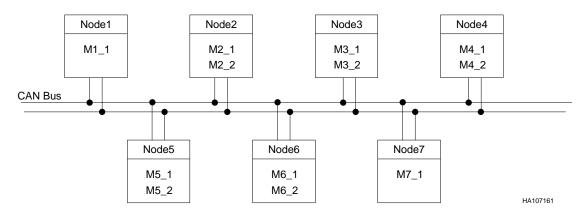


Fig. 7 Network example

The communication matrix and information about message length, repetition time and node numbers are given in table 10, whereby the described communication of the example with the according messages results in a total bus load of about 23%.

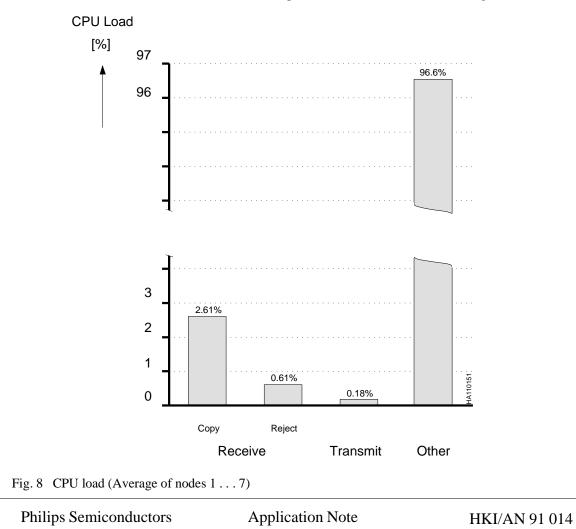
The CPU load for CAN communication is defined as the percentage of time the CPU is serving the CAN communication; its calculation is described in detail in chapter 4.6.

The diagram of figure 8 shows the CPU load of nodes 1...7, split up in receive and transmit actions of the CPU. Receive actions are executed for all messages that pass the acceptance filter. These messages are **copied** into the CPU RAM except they are not of interest for the CPU, the latter means they are **rejected**. The result of this simulation is that the CPU spends only 3.4% of the time for CAN related procedures. In this example 96.6% of the time is free for other CPU activities.

Philips Semiconductors Applica	ntion Note HKI/AN 91 014
--------------------------------	--------------------------

	unsmitting		ID	Repetition			rece	eiving n	ode		
	node / nessage	DLC	ID	Time [ms]	1	2	3	4	5	6	7
1	M1_1	4	10	2	-	х	х	х	-	-	х
2 2	M2_1 M2_2	7 2	11 30	2 100	X -	-	X -	X -	X -	X -	X X
3 3	M3_1 M3_2	5 3	15 31	20 100	X -	X -	-	-	X -	-	X X
4 4	M4_1 M4_2	5 3	32 33	20 100	-	X -	-	-	-	X -	X X
5 5	M5_1 M5_2	8 8	20 34	100 100	X -	X -	X -	-	-	X -	X X
6 6	M6_1 M6_2	8 8	21 35	100 100	-	X -	-	X -	X -	-	x x
7	M71	8	36	100	-	-	-	-	х	х	-

Table 10: Communication Matrix of the network example. (ID = Identifier, DLC = data length code)



4.6 Calculation of the CPU Load for CAN Communication

In chapter 4.5 a practical example was presented for which the CPU load for CAN communication had been investigated. The figures were obtained for a real set-up using 7 nodes. In this chapter the details of such a CPU load calculation are presented.

The following calculations and diagrams give hints and ideas to the user of the P8xC592 about the receive interrupt service execution time and CPU load when serving the CAN controller. To get results as reliable as possible all figures are calculated for worst case. As shown in fig.8 both reception of messages and transmission of messages result in CPU load, but the partial CPU load for transmission obviously is almost negligible. Therefore the considerations in the next sections have been focussed on the reception of messages is normally initiated by interrupt. The according interrupt service routine copies the Rx buffer contents to an internal CPU RAM location. Its execution time should be as short as possible in order to achieve a low CPU load.

The **CPU load** for CAN communication is defined as the percentage of time the CPU is serving the CAN communication, measured in a time interval of interest.

In case the time interval is shorter than the duration of a message transfer on the bus, the Spare Time should be considered instead of the CPU load. The **Spare Time** is defined as the amount of coherent CPU time the CPU may reserve for non-CAN activities between the reception of consecutive messages.

The next section presents the software which meets the requirements for a fast execution of the receive interrupt service.

Receive Interrupt Service Routine

Base of all the following calculations is a simple receive interrupt service routine for the P8xC592 (table 11, see also Appendix A.1). The calculations of CPU load have been investigated by using a reference software (fig.9, Appendix A.1) that uses the receive interrupt service routine. It also checks the data overrun condition in order to evaluate the Spare Time. The described software and the results are an example and transformation to user specific requirements is very easy.

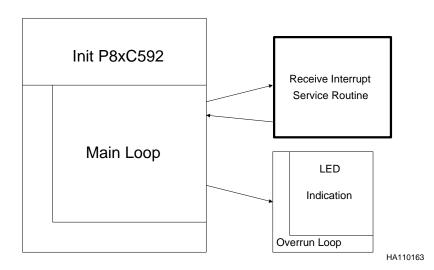


Fig. 9 Reference software with Receive Interrupt Service Routine

```
Philips SemiconductorsApplication NoteHKI/AN 91 014
```

First of all the arrangement of the reference software is given in a short summary:

Identifier:	ID10,ID9,ID8,ID7 are fixed by acceptance filter,e.g. ID = 0101 XXXX XXX
Interrupts:	Receive Interrupt enabled
Data Copy:	data bytes according data length code and a part of the identifier (as an indication for a new message) are copied into the internal μC Main RAM
Remote Frames:	in case of remote frame reception the data length code is deleted and only a part of the identifier is copied as an indication of its reception.
Overrun:	the overrun status bit is checked in the main loop. Whenever an overrun condition would be detected, indication would be done via LED. The program remains in the overrun loop until reset.
Reject:	Messages which pass the acceptance filter, but are useless for the μ C, are rejected by the software automatically.

CAN_INTERRUPT:
context switch
save Accu & PSW
select register bank
clear interrupt bits (read Interrupt Register)
calculate table address
fetch target address from table
if target address $\neq 0$, then
copy Rx Buffer to int. μ C RAM starting at target address
release Receive Buffer
(Acknowledge)
restore context

CAN_INTERRUPT:

Table 11: Flowchart of the Receive Interrupt Service Routine

Receive Interrupt Execution Time

Table 12 shows the interrupt execution time for the used receive interrupt service routine. The routine is split up into different macros, for each of them the according processing time has been calculated.

The receive interrupt execution time is shorter, when a message has to be rejected. Therefore two columns are shown, the particular processing time for the copy and the reject case. Another column shows the time passing until the Receive Buffer is released, for evaluation of the CPU Spare Time (see chapter 4.6.2).

Times are given in units of machine cycles and one cycle consists of 12 oscillator periods. The used oscillator frequency of the P8xC592 is 16 MHz resulting in a cycle time of 0.75μ s.

Fig. 10 shows the receive interrupt execution time ("copy" and "reject") as a function of the message data length code (DLC). For purpose of comparison also the message transfer time on the bus is given for the data rates: 250kbit/s, 500kbit/s and 1Mbit/s.

Philips Semiconductors	Application Note	HKI/AN 91 014
------------------------	------------------	---------------

Macro Name	Machine Cycles	Сору	Reject	Rx Buff Release
Max. Interrupt Response Time	5	Х	Х	Х
Long Jump to Interrupt Service Routine	2	Х	Х	Х
Context Switch	6	Х	Х	Х
Clear Int. Bits	1	Х	Х	Х
Calculate Target Address	15	Х	Х	Х
Copy rx buffer	8	Х	-	Х
Acknowledge	2	Х	Х	Х
Restore Context	6	Х	Х	-
Sum of cycles		45	37	39

Table 12: Summary of receive interrupt execution times (machine cycles)

Fig. 10 shows that due to the DMA function of the P8xC592 the receive interrupt execution time is always independent of the length of the received message. It is very important that the receive interrupt execution time is much shorter than the time the message itself spends on the bus. This clearly shows that a data overrun caused by a burst of messages can be excluded.

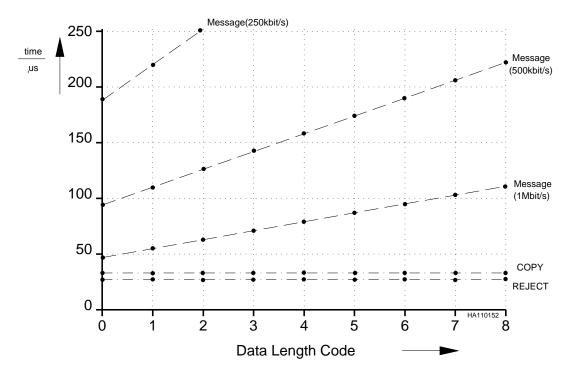


Fig.10 Receive interrupt execution time ("copy" and "reject") and message transfer time ("Message") for 16MHz CPU as a function of Data Length Code

Philips Semiconductors	Application Note	HKI/AN 91 014
1	11	

4.6.1 CPU Load

Three parameters are important for the calculation of the CPU load while receiving messages: The used receive interrupt execution time (see tables 11/12), the bit rate and the bus load.

The **bus load** is the percentage of time the bus is occupied (active) by transferring messages. Therefore it has to be calculated how many messages are transferred in a time interval, and how long a message transfer is on the bus. The latter is the message length (=number of bits) multiplied by the time of one bit. The message length depends on the data length code. For the following calculations a data length code estimation has to be made.

 $bus \ load = \frac{bus \ active \ time}{time \ interval} = \frac{number \ of \ messages \ \cdot \ message \ length \ \cdot \ bit \ time}{time \ interval}$

In order to get a direct relation between CPU Load, bit rate and bus load, from the formula for the bus load the formula for the number of messages is derived, which can be inserted into the definition of the CPU load:

number of messages = $\frac{bus \ load \cdot time \ interval}{message \ length \cdot bit \ time}$

CPU load = $\frac{number of messages \cdot receive interrupt execution time}{time interval}$

 $CPU \ Load = \frac{bus \ load \cdot time \ interval \cdot receive \ interrupt \ execution \ time}{message \ length \cdot bit \ time \cdot time \ interval}$

With the definition of *bit rate* = 1 / bit time we get

 $CPU \ load = bus \ load \cdot bit \ rate \ \cdot \ \frac{receive \ interrupt \ execution \ time}{message \ length} = bus \ load \cdot bit \ rate \ \cdot \ \tau$

The result is, that the CPU load is proportional to the bus load and bit rate. The term $\tau := receive interrupt$ execution time / message length only depends on the data length code, as it is shown in fig. 11.

Using this formula the CPU load can be calculated for any bit rate at any bus load: for example, a bus load of 10% at 400kbit/s with only one-byte messages (copy all of them) means 10% \cdot 0.4 Mbit/s \cdot 0.6 µs/bit = 2.4% of CPU load. For the example of 4-byte messages (DLC=4) the CPU load has been calculated for different bit rates and different bus loads, the result is shown in table 13.

Philips Semiconductors	
------------------------	--

Application Note

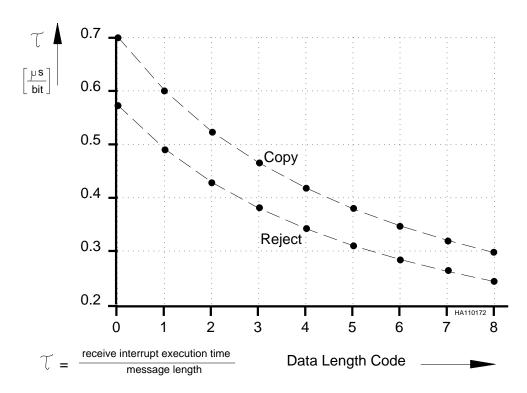


Fig. 11 Diagram for calculation of the CPU load while receiving messages

bit rate	10 % bus load	20 % bus load	50 % bus load
100 kbit/s	0.4 %	0.8 %	2.1 %
250 kbit/s	1.0 %	2.1 %	5.2 %
500 kbit/s	2.1 %	4.2 %	10 %
1 Mbit/s	4.2 %	8.4 %	21 %

Table 13: CPU load when 4-byte messages (DLC=4) are used, for worst case (copy all of them)

4.6.2 Spare Time

But, what happens when the CAN interrupt is delayed or interrupted (e.g. by other interrupt service routines with same or higher priority)? The first statement is, that the CAN receive interrupt can be delayed for at least that amount of time that is the difference between message transfer time and the receive interrupt execution time.

The actual time that a CAN receive interrupt service may be delayed without risk of loosing a message ("data overrun") is even longer: In fig.12 the execution of the receive interrupt service routine has been delayed so that the routine is not yet finished when the next message comes in, but the 'Release Receive Buffer' command comes just in time before the first byte of the next message is to be stored in the Receive Buffer. No message is lost, but of course that long delay must be compensated by faster response for subsequent reception.

Philips Semiconductors	Application Note	HKI/AN 91 014
1 milps Semiconductors	Application Note	$\Pi \mathbf{K} \mathbf{I} / \mathbf{A} \mathbf{N} 9 \mathbf{I} 0 14$

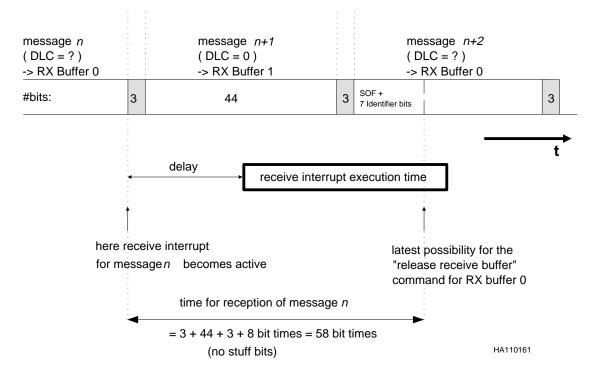
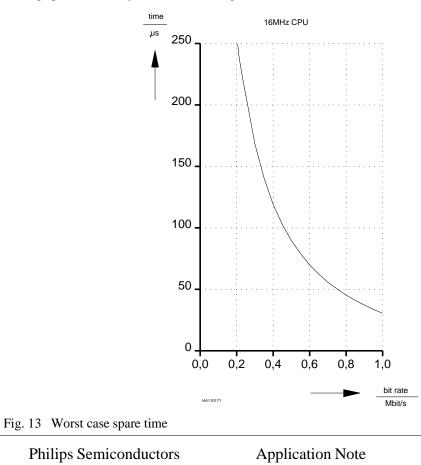


Fig. 12 Delay of receive interrupt service. The worst case is, when a message is followed by a 0-byte message plus another message of any DLC.

The resulting allowed spare time (worst case) is shown in fig.13 for different bit rates. It has been calculated by relating the Receive Buffer Release time of table 12 (39 cycles) to the transmission time of a 0-byte message plus the first byte of a next message (58 bits).



HKI/AN 91 014

5. Development Tools

For the P8xC592 several powerful support tools can be supplied to assist during the design and test phase. These tools are:

- o the P8xC592 Evaluation Board
- o P8xC592 In-Circuit Emulators
- o P87C592 EPROM programmers
- o debug tools for CAN communication NetSim NetAna NetEmu

5.1 Philips P8xC592 Evaluation Board

The P8xC592 evaluation board (OM4239) is a most versatile aid consisting of a ready-to-use hardware and software module very similar to a real CAN bus node. The P8xC592 evaluation board can be used in a car since it has its own 5V supply. An RS232 interface allows the P8xC592 evaluation board to be connected to a terminal or to a PC with terminal emulating software.

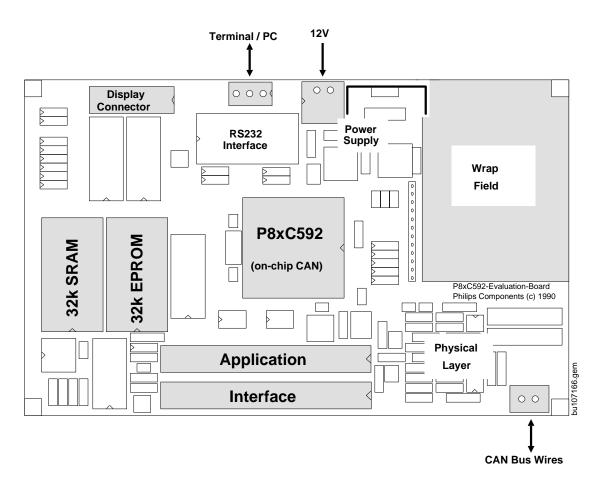


Fig. 14 P8xC592 Evaluation Board

Philips SemiconductorsApplication NoteHKI/AN 91 014

5.1.1 P8xC592 Evaluation Board Hardware

The P8xC592 evaluation board hardware (single EURO format size) is shown in figure 14. It consists of:

- P87C592 microcontroller with an external memory capacity of 32 KBytes for RAM and 32 KBytes for ROM storage. Together with the on-chip EPROM there are 48 KBytes of continuous program memory
- o 5V power supply with protection against car battery voltage fluctuations
- o off-chip transceiver circuit, designed to meet ISO/TC 22/SC 3 N608E, with two selectable options:
 - very low standby power consumption
 - support of single wire communication to recover from wiring failures.
- o RS232 interface
- o removable LED unit for demonstration purposes
- o wrap field for the user's specific circuitry

5.1.2 P8xC592 Evaluation Board Software

The software for the P8xC592 evaluation board is stored in the on-board EPROM. The software is designed to assist users with different experience in CAN-based networks:

- o A user with little experience is able to activate the demonstration software. A menu-driven software monitor allows the contents of the P87C592's CAN-registers to be altered and thereby enables the user to gain experience. For the latter an RS232-terminal (e.g. VT100) is required.
- o A user with more experience will use the "download" facility; this allows the user to load new software on to the P8xC592 evaluation board. Download requires a PC with terminal emulating software.

The P8xC592 evaluation board is also ready to be used as a bus monitor, receiving messages from the bus for display on a terminal.

5.2 In-Circuit-Emulators

For the software development on the P8xC592 several in-circuit-emulators (ICE) are available.

5.2.1 Philips Stand-alone Debug System (SDS)

Philips' well known and well accepted stand-alone debug station (SDS) for the real-time development of 8-bit μ Cs is already prepared for future probes > 16MHz. Three configurations are possible:

- o stand-alone operation together with a terminal
- o connected to a terminal plus a host computer (download/upload)
- o connected to a PC/AT (IBM or compatible) with terminal emulation software.

Philips Semiconductors	Application Note	HKI/AN 91 014
------------------------	------------------	---------------

Features of the SDS:

- o fully transparent real-time emulation
- o allows optimal emulation according to the exact specification of the target controller
- o emulation memory: 64 KBytes with no wait states
- o interfaces to external equipment
- o full assembly-level debugging
- o HLL debugging possible
- o single step and breakpoint setting
- o trace and triggering with hardware qualifiers
- o trace memory 2048 lines
- o new improved handbook
- o probes for
 - optimal connection to the target system
 - optimal emulation of the target μ C.

For emulation of the P8xC592 the user needs the following parts:

- o SDS80C51 stand-alone debug system (OM4120S)
- o probe base (OM4110)
- o emulation head "592" (OM4112)

Software for development support with the SDS (assembler, PL/M-compiler, C-compiler, HLL debugger) is offered and supported by

Tasking B.V. Plotterweg 31 3821 BB Amersfoort, Netherlands

their products can also be ordered via Philips.

5.2.2 ICE from Other Vendors

Philips supports all major third-party ICE suppliers with emulating chips for the P8xC592. Most of them already announced to support the P8xC592.

5.3 Programming Support for P87C592 (EPROM)

Philips offers a low-cost programmer for microcontrollers inclusively the P87C592:

- o LCPX5X40 programmer (OM4232)
 - 87C592-PLCC68 adaptor (OM4235)

Philips Semiconductors	Application Note	HKI/AN 91 014
------------------------	------------------	---------------

5.4 Debug Tools for CAN Communication

In addition to the P8xC592 evaluation board three other sophisticated development and debug tools for CAN communication can be supplied. These are offered and supported by

Gesellschaft für Informatik und Mikro-Elektronik Prof. Dr.-Ing. W.Lawrenz mbH Ferdinandstr. 15 A 3340 Wolfenbüttel Germany

NetSim

NetSim (Network Simulator) is a software simulator for use with a Personal Computer. The CAN network is described using NetSim on the PC by:

- o number of network nodes
- o data transmission rate
- o message identifiers / message length / message transmission repetition rate
- o noise

The simulation is then started. NetSim provides information for various parameters, such as:

- o message delays
- o bus load

NetSim assists during the design phase to investigate these parameters.

NetAna

NetAna (Network Analyzer) is a combined hardware / software tool, operating in conjunction with a PC.

NetAna has two basic functions:

- o to monitor the bus traffic and store the data on the hard disk in the PC for subsequent analysis
- o an event is triggered (identifier, bus error, etc.) and the messages around the trigger point are recorded.

NetAna assists to trace communication failures in an existing CAN network

NetEmu

NetEmu (Network Emulator) enables the user to transmit defined messages into an operational CAN network. The resultant network response can then be analyzed by NetAna.

Philips Semiconductors	Application Note	HKI/AN 91 014
------------------------	------------------	---------------

6. References

- [1] CAN Specification Version 1.2, Philips Components 1990
- [2] P8xC592 target device specification, Version 2.2, Philips Export B.V. 1991
- [3] Data sheet PCA 82C200 Stand-alone CAN-controller, Philips Export B.V. 1990, 12NC: 9397 285 30011
- [4] Application of the PCA 82C200 CAN Controller (PSCC), Philips Export B.V. 1990, 12NC: 9398 373 50011
- [5] Bit Timing Parameters for CAN Networks, Application Note KIE07/91ME, Philips Components PCALH 1991
- [6] Road vehicles Serial data communication for automotive application. Part 1: Controller Area Network (CAN). ISO/DIS 11519 part 1, International Organization for Standardization 1992
- [7] Road vehicles Interchange of digital information Controller Area Network (CAN) for high speed communication. ISO/DIS 11898, International Organization for Standardization 1992
- [8] CAN bus failure management using the P8xC592 microcontroller, Application Note HKI/AN 91 020, Philips Semiconductors PCALH 1991
- [9] Single-chip 8-bit microcontrollers: PCB83C552 User Manual, Philips Export B.V. 1988, 12NC: 9398 637 90011
- [10] CAN Physical Layer Concepts for the P8xC592 Microcontroller, Application Note HKI/AN 91 027, Philips Semiconductors PCALH 1991

Appendix

A. Reference software for investigation of the CPU load

A.1 Receiver part

This module is a program including a very fast, versatile receive interrupt service routine which is able to handle up to 128 identifiers. It runs on the P8xC592 Evaluation Board (OM4239), for instance. With only little modifications it can be adapted to other applications.

The receive interrupt execution time is less than $34\mu s$ and every software reject is done within $28\mu s.$

The acceptance filter of the CAN part is set to 0101 XXXX XXX. Upon reception of a message which passes the acceptance filter, the data bytes (plus the second byte of the Rx buffer for signalling the reception of new data to the main program) are copied into the internal RAM of the P8xC592, according to a table. In this table for each message the corresponding RAM address can be found or a "0", telling that the message is not of interest and shall be rejected.

In case of an overrun condition, indication is done via LED.

	EQU EQU EQU	0H 1H 2H 3H 4H 5H 6H 7H 8H	
CANCON	EQU EQU EQU EQU	0DBH 0DAH 0D9H 0D8H	
AUTO_RX_BUFFER RX_BUFFER DMA_RX_BUFFER	EQU	34H 14H 94H	
INT_MASK	EQU	00000010B	; Rx Interrupt enabled
ES1	BIT	0 ADH	
DSEG AT 30H ARRAY_LONG: ARRAY_SHORT		9 1	; space for Rx of an 8-byte message ; space for Rx of a 0-byte message
BIT_VAR SEGMENT RSEG BIT_VAR INT_SAVE: RX_INT BIT	DS	1	
CSEG AT RESET LJMP	MAIN_PR	OG	
CSEG AT 2BH LJMP	CAN_INT	ERRUPT	

Philips Semiconductors

Application Note

MAIN SEGMENT CODE RSEG MAIN

MAIN_PROG:

MOV	SP,#7FH	; set stackpointer
CLR	EA	; disable all interrupts
SETB	ES1	; CAN Controller interrupt enabled
		-
MOV	DPTR,#4000H	
MOV	A, #OFFH	
MOVX	@DPTR,A	; LEDs - dark
MOV	ARRAY_LONG, #0	; delete RAM position
MOV	ARRAY_LONG+1,#0	, derees min poproren
MOV	int_save,#0	; delete interrupt save register
IIT_CAN:		
MOV	CANADR, #CONTROL	; reset request
MOV	CANDAT,#0000001B	
MOV	A, CANDAT	
JNB	-	; repeat reset, if reset request is
		; absent
MOV	CANADR, #ACCEPTANCE_MASK	
MOV	CANADR, #ACCEPTANCE_MASK CANDAT, #0FH	
MOV	CANADR, #ACCEPTANCE_CODE	; TD = 0101 XXXX XXX
MOV	CANDAT,#05FH	
MOV	CANADR, #BUS_TIMING_0	
MOV	CANDAT, #80H	
MOV	CANADR, #BUS_TIMING_1	
MOV	CANDAT, #23H	
MOV	CANADR, #OUTPUT_CONTROL	
MOV	CANDAT, #0FAH	
MOV	CANADR, #CONTROL	
MOV	A, CANDAT	
ORL		; enable receive interrupt
ANL		; reset request bit = 0
MOV	CANDAT, A	
SETB	EA	; enable all interrupts
)OP:	CANSTA.1, OVERRUN	
JB	-	
SJMP	LOOP	
/ERRUN:		
MOV	A, ARRAY_LONG	
140 V	DPTR,#4100H	
MOV		; first byte of message to green Li
	@DPTR,A	, TITEL DYCE OF MESSAGE CO GIEEN DI
MOV MOVX		, TITSE Dyte of message to green h.
MOV	A, ARRAY_LONG+1	, TITST Dyte of message to green h.
MOV MOVX MOV		
MOV MOVX MOV MOV	A,ARRAY_LONG+1 DPTR,#4200H	; second byte of message to red LEI

Philips Semiconductors

Application Note

HKI/AN 91 014

------ Receive Interrupt Service Routine ------CAN_INTERRUPT: ----- CONTEXT_SWITCH ------ 6 cycles -PUSH ACC PUSH PSW USING 1 SETB RS0 CLR RS1 ----- CLEAR_INTERRUPT ----- 1 cycle -CLR_INT: MOV A,CANCON ; clear interrupt bits ----- RX_INT_TARGET_ADDR ------ 15 cycles -BEGIN RX: CANADR, #AUTO_RX_BUFFER ; set auto incr. and CANA = 20dec MOV MOV A, CANDAT ; fetch RX buffer ; upper 4 bits masked per ANL A,#0FH ; acceptance filter MOV R1,A MOV A,CANDAT ; fetch RX buffer+1 MOV R5,A ; temp. save of RX buffer+1 ANL A,#0E0H ; ID2, ID1, ID0, 0, 0, 0, 0, 0 RR ; 0,ID2,ID1,ID0,0,0,0,0 Α ORL A,R1 ; 0,ID2,ID1,ID0,ID6,ID5,ID4,ID3 ADD A, #TABLE-TABLE_BASE MOVC ; accu contains the target address A,@A+PC TABLE BASE: RELEASE_BUF ; software reject - no data copying JΖ ----- RX_INT_COPY_DATA_LOOP_DMA ------ 8 cycles -MOV CANSTA,A ; target address for DMA MOV ; get RX Buffer+1 A,R5 JNB ACC.4,DMA_START ; test of RTR bit MOV CANADR, #RX_BUFFER+1 ; in case of a Remote Frame: ANL CANDAT, #0F0H ; delete data length code, ; only the descriptor byte is copied DMA_START: MOV CANADR, #DMA_RX_BUFFER+1 NOP ; wait for data transfer NOP ----- RX_INT_ACKN ------ 2 cycles -RELEASE_BUF: CANCON, #4 ; release receive buffer MOV ----- RESTORE_CONTEXT ------ (6) cycles -END_RX: POP PSW POP ACC RETI ----- RX_INT_TARGET_ADDRESS_TABLE ------; copy data of message 01010101000 to ARRAY_LONG (8+1 bytes)
; copy data of message 01010101010 to ARRAY_SHORT (0+1 byte) TABLE: DB 0,0,0,0,0,ARRAY_LONG,0,0,0,0,0,0,0,0,0,0,0,0; ID2=0,ID1=0,ID0=0 0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0 DB ; ID2=0,ID1=0,ID0=1 DB 0,0,0,0,0,0,ARRAY_SHORT,0,0,0,0,0,0,0,0,0,0,0; ID2=0,ID1=1,ID0=0 DB 0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0 ; ID2=0,ID1=1,ID0=1 DB 0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0 ; ID2=1,ID1=0,ID0=0 DB 0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0 ; ID2=1,ID1=0,ID0=1 DB 0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0 ; ID2=1,ID1=1,ID0=0 0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0,0 DB ; ID2=1,ID1=1,ID0=1 END Philips Semiconductors Application Note HKI/AN 91 014

A.2 Transmitter part

		Tx Proced	ure	
This module is a program including a transmit procedure using a loop that copies data from the internal μ C RAM to the Tx buffer of the P8xC592. It runs on the P8xC592 Evaluation Board, for instance. Whenever the controller gets bus access, the contents of the buffer is transmitted on the CAN bus. Therefore it generates a very high bus load. The Identifier of the message has been chosen such, that it will be received by the reference software (see Appendix A.1).				
STATUS INTERRUPT ACCEPTANCE_CODE ACCEPTANCE_MASK BUS_TIMING_0 BUS_TIMING_1 OUTPUT_CONTROL CANADR CANDAT	EQU EQU EQU EQU EQU EQU EQU	0H 1H 2H 3H 4H 5H 6H 7H 8H 0DBH 0DBH 0D9H 0D9H 0D8H 8AH		
CSEG AT RESET LJMP	MAIN_PR	OG		
MAIN SEGMENT CO RSEG MAIN	DE			
MAIN_PROG:				
		Initializa	tion	
MOV	SP,#7FH		; set stackpointer	
CLR	EA		; disable all interrupts	
MOV MOV MOVX	DPTR,#4 A,#0FFH @DPTR,A		; LEDs - dark	
MOV MOV				
MOV MOV MOV MOV MOV MOV MOV		,#022H ,#033H ,#044H ,#055H ,#066H	<pre>; ID10,ID9,ID8,ID7,ID6,ID5,ID4,ID3 ; ID2,ID1,ID0,RTR,DLC3,DLC2, ; DLC1,DLC0 ; 1st data byte ; 2nd data byte ; ; ; ; ; ; ;</pre>	

INIT	CAN:			
	MOV MOV	CANADR, #CONTROL CANDAT, #0000001B	;	reset request
	MOV	A, CANDAT		
	JNB	ACC.0,INIT_CAN		repeat reset, if reset request is absent
	MOV	CANADR, #ACCEPTANCE_MASK	;	acceptance filter not used for Tx
	MOV	CANDAT,#0FFH		
	MOV	CANADR, #BUS_TIMING_0		
	MOV	CANDAT,#80H		
	MOV	CANADR, #BUS_TIMING_1		
	MOV	CANDAT,#23H		
	MOV	CANADR, #OUTPUT_CONTROL		
	MOV	CANDAT,#0FAH		
	MOV	CANADR, #CONTROL		
	MOV	A, CANDAT		
	ANL	A,#11111110B	;	reset request bit = 0
	MOV	CANDAT, A		
		Tx Loop		
BEGI	N_TX:			
	MOV	CANSTA, #ARRAY	;	target RAM address
WAIT	:			
	JNB	CANSTA.2,WAIT	;	wait until Transmit Buffer access
	MOV	CANADR, #DMA_TX_BUFFER	;	Transmit Buffer = 10H
	NOP		;	set DMA bit
	NOP			
	MOV	CANCON, #0000001B	;	transmission request
	SJMP	BEGIN TX		
	JOHE	DEGIN_IA		

END

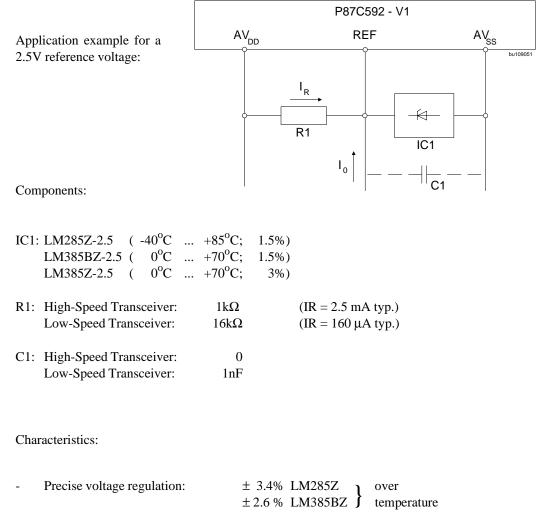
Philips Semiconductors Application Note HI
--

Specification Differences for the P87C592 - V1

Control Register / REF pin:

The bit "Reference Active" (bit 5 of the Control Register) is not provided in the P87C592-V1. This means it is not necessary to set the input function of pin REF (see Initialization procedure, chapter 4.3.1).

In the case that the REF pin (pin 55) shall be used in the application for tolerating physical bus errors - opens and shorts of the bus wires - a simple external circuit has to be connected to it.



-	Low current consumption:	(2.5 mA resp. 160 µA)

- Supply voltage range: 4.5 V to 5.5 V
- Load capacity:

Io	High-speed-circuit	Low-speed-circuit
source	-1.9 mA	-70 µA
sink	5 mA	5 mA

Philips Semiconductors

1991-10-16